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PATENT APPLICATION  
Attorney Docket No.: 678-1217 (P10800)

the minimum coding gain becomes flat. Therefore, a phase value preferable in the first embodiment of the invention is 45°. FIG. 6 illustrates a QPSK constellation which is phase-rotated by 45°. As illustrated, the phase-rotated symbols are situated on a real axis or an imaginary axis. According to the first embodiment of the invention, a preferable phase rotation range is between 21° and 69° centering on 45° for QPSK, between 21° and 24° for 8PSK, and is 11.25° for 16PSK, ~~centering on 45°~~. However, the invention is not restricted to the figures, and the preferable phase rotation range shall be set according to characteristics of the system.---

Please replace the second full paragraph on page 16, beginning at line 16, with the following:

---If a metric value is calculated with channel gains  $h_1$ ,  $h_2$  and  $h_3$  from 3 transmission antennas to a reception antenna for Equation (14), it becomes

$$\frac{|r_1 - h_1 e^{-j\theta_1} s_1 - h_2 e^{-j\theta_2} s_2 - h_3 s_3|^2 + |r_2 - h_1 s_3 - h_2 e^{-j\theta_1} s_1 - h_3 e^{-j\theta_2} s_2|^2}{|r_3 - h_1 e^{-j\theta_1} s_2 - h_2 s_3 - h_3 e^{-j\theta_1} s_1|^2}$$

$$\frac{|r_1 - h_1 e^{-j\theta_1} s_1 - h_2 e^{-j\theta_2} s_2 - h_3 s_3|^2 + |r_2 - h_1 s_3 - h_2 e^{-j\theta_1} s_1 - h_3 e^{-j\theta_2} s_2|^2}{|r_3 - h_1 e^{-j\theta_1} s_2 - h_2 s_3 - h_3 e^{-j\theta_1} s_1|^2} \quad \dots(15)$$

A receiver then determines symbols  $s_1$  to  $s_3$  that minimize Equation (15). ---

Please replace the first paragraph on page 17, beginning at line 4, with the following:

--- FIG. 8 is a block diagram illustrating a structure of a transmitter using a space-time block code according to a second embodiment of the present invention. As illustrated, the receiver

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12/20/07

$$\begin{pmatrix} e^{j\theta_1} s_1 & s_2 & e^{j\theta_4} s_4 \\ -s_2^* & e^{-j\theta_1} s_1^* & s_3^* \\ -e^{-j\theta_4} s_4^* & -s_3^* & e^{-j\theta_1} s_1^* \\ s_3 & -e^{j\theta_4} s_4 & s_2 \end{pmatrix}$$

.....(11)

Equation (11) shows an encoding matrix for phase-rotating  $s_1$  and  $s_4$  among input symbols  $s_1, s_2, s_3$  and  $s_4$  of Equation (7) by  $\theta_1$  and  $\theta_2$ , respectively. In another case, it is possible to rotate a symbol pair of  $(s_1, s_2), (s_3, s_4)$  or  $(s_2, s_3)$  related to different matrixes. Although phase values by which the 2 symbols are rotated respectively are different from or identical to each other, a diversity order is always maintained at 3. Likewise, if 2 symbols that determine different metric values are phase-rotated by a predetermined phase value even for the other encoding matrixes of Equation (8), final encoding matrixes can be obtained. ---

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12/21/07

Please replace the third full paragraph on page 12, beginning at line 21, with the following:

--- For example, when  $s_1$  and  $s_4$  among 4 input symbols  $s_1, s_2, s_3$  and  $s_4$  are phase-rotated by  $\theta_1$  and  $\theta_2$ , respectively, an output of the encoder 230 can be expressed in a  $4 \times 3$  encoding matrix of Equation (11) above. When the encoding matrix of Equation (11) is used, 3 symbols  $e^{j\theta_1} s_1, s_2$  and  $e^{j\theta_4} s_4$  in a first row are delivered to the 3 antennas 240, 242 and 244, respectively, in a first time interval and symbols  $s_3, e^{-j\theta_4} s_4$  and  $s_2$  in the last 4<sup>th</sup> row are delivered to the 3 antennas 240, 242 and 244, respectively, in the last 4<sup>th</sup> time interval.---

Please replace the third full paragraph on page 14, beginning at line 18, with the following:

---It can be understood from the result of FIG. 5 that when all phase values exist at around  $45^\circ$ ,